

# A time-series method to identify and correct range sidelobes in meteorological radar data

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## ABSTRACT

The use of pulse compression techniques to improve the sensitivity of meteorological radars has become increasingly common in recent years. An unavoidable side-effect of such techniques is the formation of ‘range sidelobes’ which lead to spreading of information across several range gates. These artefacts are particularly troublesome in regions where there is a sharp gradient in the power backscattered to the antenna as a function of range.

In this article we present a simple method for identifying and correcting range sidelobe artefacts. We make use of the fact that meteorological targets produce an echo which fluctuates at random, and that this echo, like a fingerprint, is unique to each range gate. By cross-correlating the echo time series from pairs of gates therefore we can identify whether information from one gate has spread into another, and hence flag regions of contamination. In addition we show that the correlation coefficients contain quantitative information about the fraction of power leaked from one range gate to another, and we propose a simple algorithm to correct the corrupted reflectivity profile.

## 1. Introduction

Pulse compression is a popular method to increase the sensitivity and/or range-resolution of meteorological radars. In recent years pulse compression has found extensive use in millimeter cloud radars (e.g. Kollias *et al* 2007; Moran *et al* 1998) in order to detect the weak echoes associated with thin stratocumulus or cirrus clouds where the particles are small. Pulse compression is commonly used by Mesosphere-Stratosphere-Troposphere wind profilers (e.g. Hooper *et al* 2008), and it has also been proposed as a means to increase the number of independent samples for rain radars with short dwell times (Mudukutore *et al* 1998).

A conventional pulsed radar has a sensitivity which is limited by the peak power output of the transmitter multiplied by the length of the pulse (in addition to other factors such as antenna size, dwell time, etc). Longer pulses lead to higher sensitivity, but also to poorer range-resolution. Pulse compression attempts to improve the sensitivity of a radar, while maintaining high range resolution. This is achieved by transmitting a long pulse which has extra information encoded into it on time scales corresponding to the desired range resolution. The echo is then decoded using a matched filter. The encoded information takes the form of either phase or frequency modulation - see Farnett and Stevens (1990) for a review of the various implementations which are possible.

A side-effect of pulse compression is the formation of

range sidelobes, where echoes from a given range leak into neighbouring range gates. In essence this occurs where there is not enough information encoded into the transmitted pulse to uniquely decode the reflected long pulse into the desired short range resolution, or where that encoded information has been corrupted by the motion of the particles being probed.

Because these sidelobes are typically much smaller in magnitude than the echo from which they originate, in regions where the power backscattered to the antenna is quite uniform with range the sidelobes have little effect on quantities of interest such as the radar reflectivity or Doppler velocity spectrum. However, where there are sharp gradients as a function of range, as is common in clouds and precipitation, the influence of sidelobes can be significant and problematic for quantitative interpretation of the measurements. Such artefacts are particularly important for dual-wavelength techniques where even small biases can lead to large retrieval errors (Hogan *et al* 2005).

At present there is no objective method to identify range sidelobe artefacts, or to correct the reflectivity data for their influence. Some radars interleave short uncompressed pulses between the long compressed ones, allowing a cross-check at some gates; however, since the aim of pulse compression is to detect echoes which cannot be detected using an uncompressed pulse at the same range resolution, artefacts cannot be diagnosed this way in many cases. Em-

pirical methods of flagging affected data, based on identification of situations likely to give rise to artefacts (eg where there are sharp gradients in backscatter with range) have been developed (e.g. Moran *et al* 1998) - however it is highly desirable to develop more rigorous techniques.

In this article we develop a new idea to identify and correct for the presence of range sidelobes. We show that the cross correlation between echo time series sampled at pairs of range gates contains information on the occurrence of sidelobes. This is first used to visualise the locations of sidelobe artefacts in a drizzling stratus cloud profile, and then to flag the gates corrupted by sidelobes objectively. Next, we show that there is a quantitative link between the correlation coefficients and the leakage of power from one range gate to another, and we use this fact to develop a tentative algorithm for correcting a reflectivity profile corrupted by range sidelobes. We test these ideas on a second example (a mid-level ice cloud) and conclude with a brief discussion and directions for future work.

## 2. Method

The essential idea which we will utilise in this study is that meteorological targets produce an echo which fluctuates from pulse to pulse (Marshall and Hitschfeld 1953; Wallace 1953). This fluctuation occurs as particles move relative to one another on scales of order one quarter of the wavelength, leading to waves which may add constructively or destructively at the radar antenna. Since this reshuffling occurs on scales which are very small compared to the range resolution of the radar (a factor  $\sim 10^4$  difference in scale for the radar in section 3), this leads to the expectation that each range gate will sample a different fluctuating echo to any other range gate (in other words, each time series is *unique*). This means that for a pair of range gates  $i \neq j$  we anticipate that the correlation coefficient  $\rho_{ij}$  between the time series samples at those two gates should be zero, since the time series are uncorrelated with one another. In practice this is only true in the limit where the length of the time series is much longer than the decorrelation time of the echo, and for a finite time series this means that  $|\rho_{ij}|$  will be slightly greater than zero. However we expect that it should be rather small for many radar configurations, and this expectation is verified observationally in sections 3 and 4.

Where range sidelobes are present, we should expect  $|\rho_{i \neq j}|$  to be significantly greater than zero. This is because information has leaked from a range gate where the echo is strong, to another range gate where the echo is weak, giving rise to a correlation between the two gates. To detect range sidelobe artefacts therefore, we simply calculate  $|\rho_{ij}|$  and look for values significantly above zero.

## 3. Example: drizzling stratus

We illustrate the method outlined above using data from the National Centre for Atmospheric Science (NCAS) 35GHz cloud radar at the Chilbolton Observatory in the UK. We are interested in the general method rather than the specifics of the pulse compression scheme used, and so we describe the details of the compression in brief only.

Two complementary 10-bit binary phase codes are transmitted in sequence. Complementary codes have the advantageous property that the range sidelobes produced by each code theoretically cancel each other out when the decoded signals are summed. In practice (and in the example shown here) this is not always achieved because the assumption that the echo is unchanged between the first and second coded pulse is frequently violated by meteorological scatterers moving relative to the antenna, and relative to one another, by a significant fraction of the Nyquist velocity (Wakasugi and Fukao 1985): hence range sidelobes are produced. The length of the coded pulses is  $4 \mu\text{s}$ , leading to a compressed range resolution of 60m for our 10-bit code. Coded pulses are transmitted every 0.2 ms. The data is oversampled in range at intervals of 30m.

As well as long coded pulses, short  $0.4 \mu\text{s}$  pulses with no compression are interleaved between them. Again the range resolution is 60m oversampled to 30m and the pulses are transmitted every 0.2 ms. The coded measurements are approximately 13 dB more sensitive than the uncoded pulse data as a consequence of the longer pulse, coherent averaging of the two complementary coded pulses, and the oversampling in range by the receiver.

On 28 December 2012 a layer of drizzling stratus cloud was observed over much of southern England. Figure 1 shows two hours of radar reflectivity data measured using the radar at Chilbolton whilst dwelling at vertical. Figure 1a shows the reflectivity measured using the uncoded pulses, whilst figure 1(b) shows the results measured using the coded (ie compressed) pulses. While both methods yield similar reflectivity fields at lower levels, there are weak echoes in the coded data above  $\approx 2\text{km}$  which are not present in the uncoded data. This behaviour has been observed quite frequently in drizzling boundary layer clouds, and we strongly suspect that these are range sidelobes resulting from the pulse compression.

In what follows we will test this hypothesis objectively using a 0.5 s sample of echo time series data collected at 2200 UTC (indicated by the dashed line in figure 1). This corresponds to 2048 coded and uncoded pulses (note however since pairs of coded pulses are combined, this leads to a coded pulse time series which is only 1024 points in length). Figure 2 shows the specific profile being considered in detail: here the signal to noise ratio (SNR)<sup>1</sup> is shown as

<sup>1</sup>We define  $\text{SNR} = (P - \mu) / \sigma$  where  $P$  is the received power,  $\mu$  is the mean noise power sampled in empty (noise-dominated) range gates,

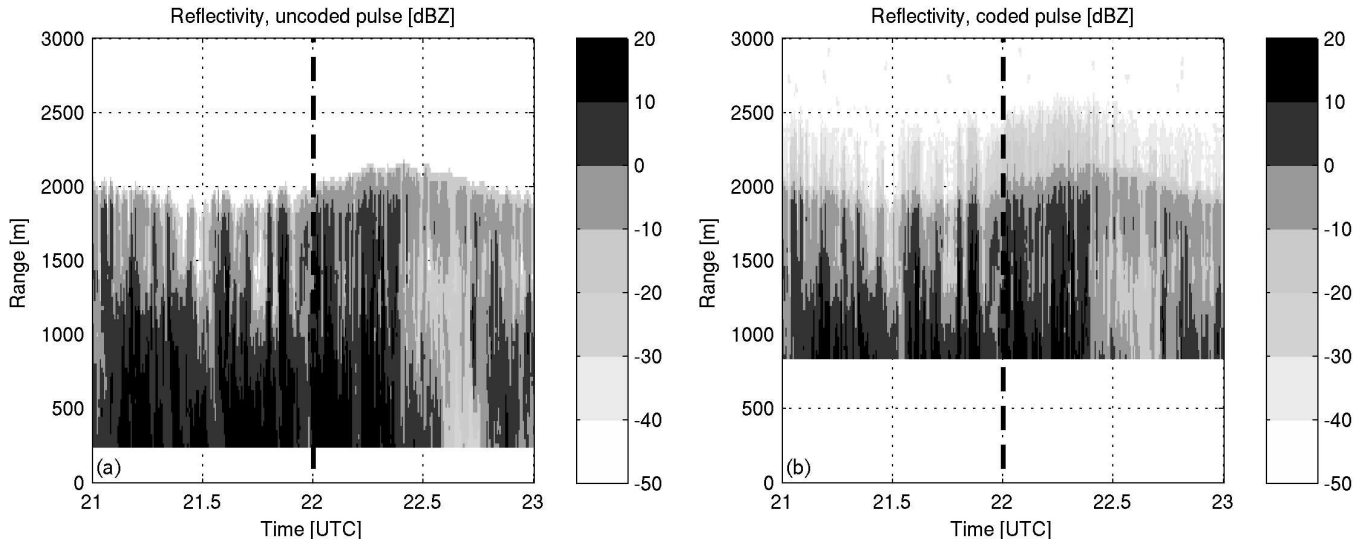


FIG. 1. Radar reflectivity measured through a drizzling stratus cloud over a two-hour period while the radar was dwelling at vertical. Panel (a) shows data collected using a simple uncoded pulse; panel (b) shows data collected using a 10-bit complementary-pair coded pulse (see text). Note the weak echoes above  $\approx 2000\text{m}$  in panel (b) which are absent from panel (a): these are suspected to be range sidelobes. There is a small dead time after transmission of the pulse during which no echoes are received: for the uncoded data in panel (a) this blind zone is 200m; for the coded data in panel (b) this is augmented by the length of the long coded pulse leading to a larger blind zone of 800m. The dashed line indicates the vertical profile which will be investigated in more detail in sections 3a,b and 4.

a function of range for both coded and uncoded pulses. Note SNR drops to 0dB for the uncoded pulse at 1940m; meanwhile the coded pulse detects a significant echo up to 2360m. Also shown for reference is the uncoded SNR with a 13dB offset (dashed line). At ranges  $< 1700\text{m}$  the profiles are almost identical. Between 1700–1940m however the gradient of SNR with range is steep ( $\sim 6\text{dB}$  per 100m), and here the coded pulse overestimates the reflectivity relative to the uncoded data: at 1820m this difference is 1dB; while at 1880m the difference is 4dB, suggesting that range sidelobes may be affecting these higher altitude gates.

#### a. Correlations between pairs of time series

The correlation coefficient  $\rho_{ij}$  between the complex samples at each possible pair of range gates was computed for both coded and uncoded pulses. Note that in an operational algorithm one would only need to compare gates within one code length from each other - here we show all possible correlations for completeness. Figure 3 shows the magnitude of the correlation coefficient (grey scale) as a function of range on the ordinate and abscissa. Note that the diagram is symmetrical, since  $|\rho_{ij}| = |\rho_{ji}|$ . In figure 3a, the uncoded echo time series are, as expected, uncorrelated with one another and values of  $|\rho_{ij}|$  are small (mean value of  $|\rho| = 0.05$  in regions of  $\text{SNR} > 5\text{dB}$ , i.e. where  $\sigma$  is the standard deviation of that noise from gate to gate.

the sample is dominated by meteorological echoes rather than noise). The exception to this rule is the diagonal elements of the figure, where  $i = j$  and one is simply correlating the time series at that gate with itself ( $|\rho_{ii}| = 1$ ). Note also that because of the oversampling to 30m bins, there is some correlation between neighbouring bins where a detectable meteorological echo is present (800–1900m in the figure). At ranges where noise is dominating the echo ( $> 1900\text{m}$ ) this correlation between neighbouring gates disappears, and  $|\rho|$  is very close to zero, demonstrating that the receiver noise is uncorrelated from gate to gate.

Figure 3b shows the same correlation coefficients for the coded pulses. The influence of the range sidelobes is immediately obvious in this figure, with correlation coefficients significantly higher than zero present in the off-diagonal elements (peak values of  $|\rho_{ij}|$  in this case were  $\approx 0.7$ ). The regions of correlation are very clearly defined, running parallel to the diagonal, but offset from it by  $\approx 100\text{m}$  and  $\approx 400\text{--}600\text{m}$  (indicated with arrows on one half of the figure). This is the behaviour which we were led to anticipate from range sidelobe artefacts in the discussion in section 2. We therefore identify these areas of correlation as range sidelobes produced by the strong drizzle echoes between 1400–1800m range masking the much weaker (or absent) echoes above. These well defined features demonstrate that the correlation coefficients can provide a robust indicator for the presence of sidelobes, which can be ex-

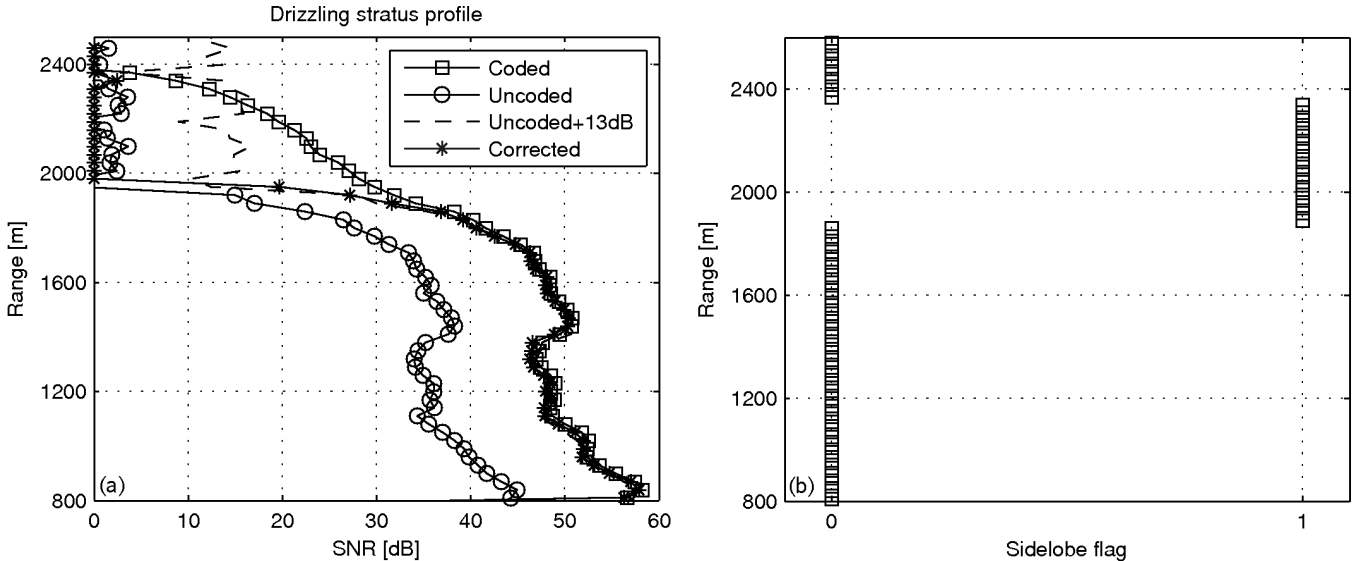


FIG. 2. Panel (a) shows the vertical profile of signal to noise ratio through the cloud. Squares are for the complementary-pair coded pulse described in the text. Circles show the same profile measured using a simple uncoded pulse. The dashed line is simply the uncoded profile shifted by 13dB to aid comparison with the coded data (which is 13dB more sensitive). Panel (b) shows the sidelobe flag derived from the simple identification algorithm described in the text. A value of one indicates significant range sidelobe artefacts are likely at that range gate.

ploited to identify affected gates.

#### b. A simple flag for sidelobe artefacts

Having graphically identified those pairs of gates where information has leaked from one gate to another, we suggest a rudimentary algorithm to flag corrupted data. The implementation of a fully-developed operational algorithm to flag and remove range sidelobe artefacts is beyond the scope of this initial study. However we have experimented with some simple approaches. The most obvious idea is simply to flag any range gate where there is a significant correlation with another gate separated from it by less than the length of the coded pulse. However in the example presented here, this would remove much of the data at ranges 1400 – 1800m, even though the agreement between the coded and uncoded SNR profiles here is very close.

A simple refinement is to identify ‘which way’ the information is likely to be flowing. Specifically we assume that weak signals (or background noise) is masked by sidelobes from much stronger signals, and not vice-versa. Given a correlation coefficient between a pair of gates above some threshold value  $|\rho|_{\text{crit}}$ , one then seeks the weaker SNR among the two gates, and flags that data as affected, whilst leaving the stronger signal unflagged. The result of this simple algorithm is shown in figure 2(b) alongside the SNR profile. Here we have chosen  $|\rho|_{\text{crit}} = 0.25$  although the results were not found to be sensitive to this choice of threshold,

and identical flagging was obtained using  $|\rho|_{\text{crit}} = 0.15$  and  $|\rho|_{\text{crit}} = 0.35$ . Data at ranges between 1850–2400m have been flagged as corrupted, whilst the data at lower ranges has not. This is an encouraging result. The algorithm has removed the spurious echoes above 1900m where there is in fact almost certainly no cloud present. It has also identified a few pixels near cloud top where the gradient in SNR is steep and where we had noted discrepancies between coded and uncoded results. This gives us optimism that this relatively simple algorithm can be used to identify sidelobe artefacts.

#### 4. Correcting the corrupted reflectivity profile

To correct the corrupted reflectivity profile for the influence of range sidelobes, a quantitative link is needed between the correlation coefficients measured in section 3a and the amount of power being leaked from a range gate with a strong echo to another range gate with a weak echo. Consider the complex samples  $V$  of the echo measured at range gate  $i$ :

$$V_i^m = V_i^t + \sum_{j=i-CL}^{i+CL} f_{ji} V_j^t \quad (1)$$

where  $V_i^t$  is the true echo at range gate  $i$ , and  $V_i^m$  is the (potentially corrupted) measurement. The second term on the right hand side represents the contributions from any sidelobes originating in other range gates within one code

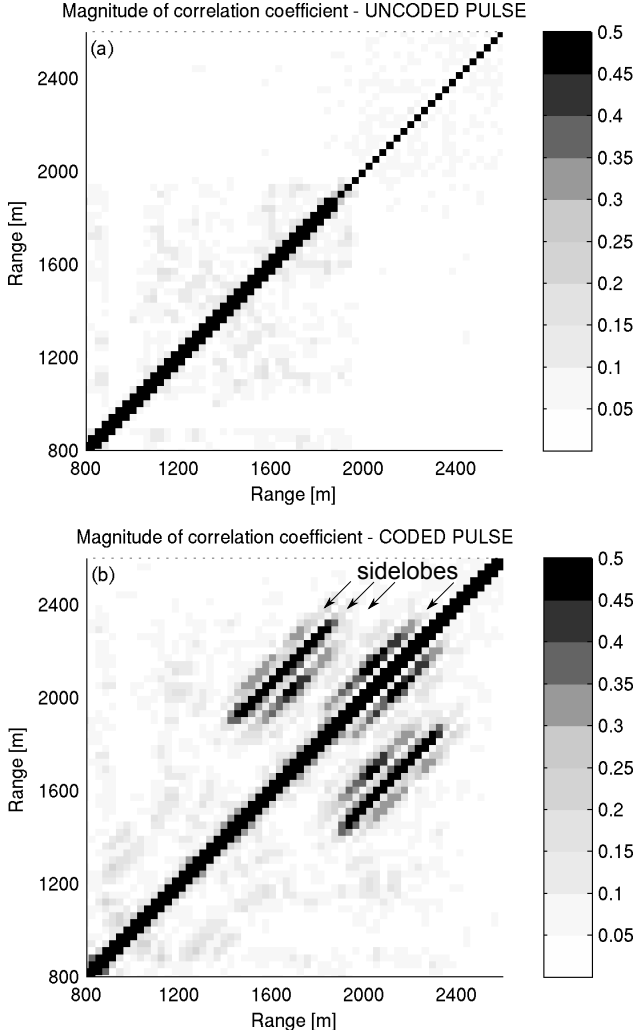


FIG. 3. Correlation coefficients computed for complex time series at each pair of range gates between 800 and 2600m. The absolute magnitude of the correlation coefficient is shown here (grey scale). Panel (a) shows results for a simple uncoded pulse; panel (b) shows results for a 10-bit complementary-pair coded pulse (see text). Note the off-diagonal regions of high correlation in panel (b): these are correlations arising from range sidelobes. The diagrams are symmetrical since  $|\rho_{ij}| = |\rho_{ji}|$ .

length  $CL$  of gate  $i$  (for our 35GHz radar  $CL = 10$ ). The factors  $f_{ji}$  represent the sidelobes themselves, specifically they are the (possibly complex) fraction of the true echo at gate  $j$  which has leaked into the measured echo at gate  $i$ . We assume in this analysis that  $|f_{ji}| \ll 1$ .

The cross correlation coefficient between a pair of measured time series at gates  $k$  and  $i$  is:

$$\rho_{ki} = \frac{\sum_{\text{time}} V_k^m (V_i^m)^*}{\left( \sum_{\text{time}} |V_k^m|^2 \sum_{\text{time}} |V_i^m|^2 \right)^{0.5}} \quad (2)$$

Substituting equation 1 into the numerator, expanding out, and neglecting terms of order  $f^2$ , one obtains three terms remaining in the numerator  $N$ :

$$N = \sum_{\text{time}} V_k^t (V_i^t)^* + \sum_{\text{time}} \left[ (V_i^t)^* \sum_{j=k-CL}^{k+CL} f_{jk} V_j^t + V_k^t \sum_{j=i-CL}^{i+CL} (f_{ji} V_j^t)^* \right] \quad (3)$$

The first term approaches zero for a sufficiently long time series, since  $V_k^t$  and  $V_i^t$  are uncorrelated. All of the terms in the first sum over  $j$  are approximately equal to zero for the same reason, except when  $j = i$ . Finally, all of the terms in the second sum over  $j$  are equal to zero, except when  $j = k$ . This yields the result:

$$\rho_{ki} = \frac{\sum_{\text{time}} f_{ik} |V_i^t|^2 + f_{ki}^* |V_k^t|^2}{\left( \sum_{\text{time}} |V_k^m|^2 \sum_{\text{time}} |V_i^m|^2 \right)^{0.5}} \quad (4)$$

Now we recognise that since  $|f| \ll 1$  we need only concern ourselves with the scenario where one gate  $k$  contains a strong echo, and the second gate  $i$  contains a much weaker echo (ie  $|V_k^t|^2 \gg |V_i^t|^2$  and  $|V_k^t|^2 \approx |V_k^m|^2$ ). We may then simplify further to obtain the magnitude of the correlation coefficient (as measured in figure 3):

$$|\rho_{ki}| = \left| \frac{\sum_{\text{time}} f_{ki} |V_k^m|^2}{\left( \sum_{\text{time}} |V_k^m|^2 \sum_{\text{time}} |V_i^m|^2 \right)^{0.5}} \right| \quad (5)$$

or equivalently:

$$|\rho_{ki}|^2 = \langle f_{ki} \rangle^2 \times \left( \frac{\sum_{\text{time}} |V_k^m|^2}{\sum_{\text{time}} |V_i^m|^2} \right) \quad (6)$$

where

$$\langle f_{ki} \rangle = \frac{\sum_{\text{time}} f_{ki} |V_k^t|^2}{\sum_{\text{time}} |V_k^t|^2} \quad (7)$$

is the time-averaged leakage factor, weighted by the power of the echo at gate  $k$ .

Equation 6 reveals how the correlation coefficient is quantitatively related to the sidelobe leakage from one gate to another. On the basis of this, we now propose a simple algorithm to determine these time-averaged leakage factors  $|\langle f_{ki} \rangle|$  and we use these to correct the profile of SNR (and

hence the radar reflectivity). Note that it is the power measured by the receiver which is to be corrected: range-weighted quantities such as radar reflectivity should only be computed after the correction is complete.

The algorithm is extremely simple. At each gate  $i$  one identifies other range gates  $k$  within one code length which have a signal to noise ratio significantly higher than that at  $i$  (here we use 3dB as a threshold). If we take the amount of power leaked from gate  $k$  to gate  $i$  to be  $|\langle f_{ki} \rangle|^2 \times \text{SNR}_k$ , then equation 6 shows that this is equal to  $|\rho_{ki}|^2 \times \text{SNR}_i$ . In other words,  $|\rho_{ki}|^2$  represents the fraction of the echo power at gate  $i$  which is introduced from the stronger return at gate  $k$ . Given this, we can use our measured correlation coefficient  $|\rho_{ki}|$  to compute the contribution to the SNR at gate  $i$  which is due to leakage from gate  $k$ . We subtract this number from the corrupted SNR value at gate  $i$  for  $k$  within one code length. This provides us with a corrected SNR value for that gate. We then move on and repeat the process for all other range gates, yielding a complete corrected SNR profile.

Figure 2a shows the result of this correction process for our drizzling stratus example (solid line with \* symbols). The spurious echoes above 2km have now been completely removed. In addition the reflectivity profile at the top of the cloud ( $\approx 1.8\text{-}2\text{km}$  range) is now in close agreement with the uncoded profile, showing a sharp decrease in reflectivity with range. This result is extremely encouraging, giving us a profile which matches the uncoded data and provides a correct cloud top height. This indicates that the procedure outlined above is a sensible one, and that the analysis above is a valid approximation to the problem.

One source of error in this approach is that it is likely to slightly overestimate the amount of power which should be subtracted from the corrupted gates. This is because even in the absence of sidelobes,  $|\rho_{ki}| > 0$  due to the finite decorrelation time of the echo relative to the total length of the time series. This ‘noise’ in  $|\rho_{ki}|^2$  can be significant, and is additive when sidelobes are present, leading to excess power being subtracted from a corrupted gate. In the gates above 2km in our example above (where we should have subtracted 100% of the signal) our algorithm actually subtracted between  $\approx 100\text{-}130\%$  in many of the gates. On the figure we have set these negative points to a value of  $\text{SNR}=0\text{dB}$ . This issue can be ameliorated by using a longer time series, which will yield a more accurate correction since the noise in  $|\rho_{ki}|^2$  will be reduced.

A second possible source of error is the form of the retrieved correction factor itself. We are estimating the leakage factor  $f_{ki}$  for the complex samples, weighted by the power in gate  $k$  on each pulse, and averaged over the time series. However, the leakage of *power* from one gate to another is proportional to  $|f_{ki}|^2$ , and therefore it is in

fact the quantity

$$\langle |f_{ki}|^2 \rangle = \frac{\sum_{\text{time}} |f_{ki}|^2 |V_k^t|^2}{\sum_{\text{time}} |V_k^t|^2} \quad (8)$$

which we seek in order to correct the reflectivity profile. In the correction procedure above we have approximated this by  $|\langle f_{ki} \rangle|^2$ . This is reasonable if  $f_{ki}$  does not vary greatly over the time series. The agreement between the corrected profile and the uncoded profile in figure 2a is evidence that this approximation is indeed an acceptable one, at least for our radar setup.

## 5. Example II: the base of a thick mid-level ice cloud

To illustrate the idea further, we now briefly present a second example profile, this time collected in a deep mid-level ice cloud sampled at 1500 UTC on 23 December 2012. The experimental set up is identical to section 3. Figure 4a shows the uncoded SNR profile: note the sharp gradient close to the cloud base between 1.7–2km range, as the ice particles fall into dry air below and evaporate. The coded profile is consistent with the uncoded measurements at ranges above 1.75km; however between  $\approx 1.65\text{-}1.75\text{km}$  the coded and uncoded measurements disagree, and between 1.45–1.65km (where the uncoded signal is dominated by noise) the coded profile contains a significant echo, which we suspect to be the result of range sidelobes.

Figure 4b shows the correlation values for all pairs of range gates between 1–2.5km using the uncoded time series. As in the stratus case, we observe  $|\rho_{i \neq j}| \approx 0$ . Figure 4c shows the same for the coded time series - this time two lines of significant correlation ( $|\rho_{i \neq j}| \approx 0.5$ ) are present, and show that the time series at range gates between 1.45–1.75 are highly correlated with the time series sampled approximately 400 and 550m higher up the profile.

Using the same algorithm described in section 3b the areas of likely contamination were automatically flagged, and this is shown in figure 4d. The algorithm has diagnosed all gates between 1450–1750m as corrupted, and this is consistent with the region where we suspected that no cloud was present. It has also flagged the region between 1.65–1.75km where the uncoded and coded profiles did not match.

Finally, we have corrected the coded SNR profile using the procedure in section 4. The corrected profile is shown in figure 4a. As in the stratus case, the corrected profile shows very close agreement with the uncoded profile, and has again removed all of the signal in gates where we suspected there was no cloud, leading to a correct cloud base height. This time between 90–110% of the echo was removed in gates where we inferred that there was no cloud present. There is one range bin at 1500m where the corrected SNR is +7dB, and where we believe cloud is absent.

However this single pixel would be easily removed with a simple speckle filter (used as part of our standard processing). These results encourage us further that our correction methodology is a reasonable one.

## 6. Conclusions and discussion

We have shown how pulse-to-pulse fluctuations can be used to diagnose the presence of range sidelobe artefacts associated with pulse compression, and have illustrated the approach using data collected from a drizzling stratus cloud and a thick mid-level ice cloud with the 35GHz cloud radar at Chilbolton. We have suggested a simple algorithm to flag the sidelobe-affected data. We have also shown how the correlation coefficients are quantitatively linked to the amount of the power in a particular range gate which originates as the result of a sidelobe from a second range gate, and hence we have formulated a simple correction algorithm which appears to perform well on the two examples presented here.

While the example data presented here was collected using complementary phase-coded pulses, the methodology is quite general and ought to be applicable to other pulse compression techniques, provided that a decoded time series of echo amplitudes is available which can be cross-correlated. We emphasise that although we have shown simultaneous uncoded measurements in figures 1–4 for the purposes of comparison, it is only the compressed pulse data which are used in the flagging and correction algorithms.

One outstanding issue is the small spurious correlations which are produced when the decorrelation of the signal is slow relative to length of time series (i.e. the number of independent samples is small). This makes it more difficult to identify some of the less prominent sidelobe artefacts, and leads to an over-correction of the reflectivity profile, which we would like to minimise, or at least quantify. Of course the magnitude of these correlations will be different depending on the meteorological situation. In cirrus clouds where the Doppler spectrum is very narrow for a vertically pointing radar, and decorrelation of the echo is corresponding slow, we have measured mean and standard deviation of  $|\rho_{i \neq j}|$  of  $\approx 0.1$  using uncoded pulses. This leads to a few pairs of range gates where the correlation is as large as 0.3 despite the absence of sidelobes. The most logical way to improve this situation is to make use of the longest possible time series, taking into account the desired time resolution of the flagged/corrected data. We are currently studying the possibility of setting a variable  $|\rho_{ij}|$  threshold in our flagging algorithm which is diagnosed based on the decorrelation time (or Doppler spectrum width) of the samples in question.

A second outstanding issue is the neglect of terms of order  $f^2$  in the derivation of equation 3. The neglected

terms are of the form  $f_l f_k^* |V_l^t|^2$ . Physically, these terms correspond to correlation between gates  $k$  and  $i$  being introduced via sidelobes from a third gate  $l$  which is leaking power into both  $k$  and  $i$ . All three gates must lie within one code length of each other. These terms become significant in the scenario where the meteorological echoes at both  $k$  and  $i$  are comparable and weak relative to the echo at gate  $l$ . This does not introduce any difficulties in the flagging algorithm, but it does emphasise the need to only apply the correction algorithm to gate pairs where  $|V_k^t|^2 \gg |V_i^t|^2$  (in the profiles shown a 3dB threshold was used, but depending on the magnitude of  $f$  a higher threshold might prove more robust). We plan to investigate the role of these higher order correlations theoretically via computer simulation (see future work below).

Another source of range sidelobes (and hence correlation) between range gates may occur in the absence of pulse compression, simply as a result of the finite bandwidth of the receiver and finite duration of the transmitted pulse, leading to a slight spreading of information across neighbouring range gates (e.g. Doviak and Zrnić 1984; Nicol and Illingworth 2013). However, in section 3a we noted that where the echo is dominated by noise the time series were uncorrelated from one range bin to the next. This is evidence that the receiver filter effect is not a significant source of correlation, at least for the 35GHz radar used here.

Unlike the reflectivity profile, we have not attempted to correct the Doppler information. This is because Doppler velocity depends on the change in phase between pairs of pulses, and hence a correction of the echoes on a pulse by pulse basis would be required. The correlation coefficient on the other hand only describes a time-averaged leakage factor  $\langle f_{ki} \rangle$ . In section 4 we suggested that  $f_{ki}$  might be approximately constant in time ( $f_{ki} \approx \langle f_{ki} \rangle$ ). However the Doppler information is much more sensitive to the accuracy of this approximation, since any over- or under-correction on individual pulse amplitudes will effectively introduce spurious power into the Doppler spectrum. More analysis is needed to investigate this aspect.

So far we have only applied our method to a vertically-pointing cloud radar. It is interesting to consider whether the same approach could be applied to scanning radars. For scanning cloud radars, or research radars which can scan slowly, it should be possible to apply the present methodology. However, for operational weather radars the dwell time per ray is usually very short, while decorrelation of the echo is relatively slow because of the longer wavelength: for example Illingworth (2003) suggests that one might expect only around 35 independent samples when scanning in rainfall at low elevations. The vertical dwells analysed in sections 3 and 5 had a spectral width of  $\approx 0.3\text{ms}^{-1}$  corresponding to 120 independent samples. According to Fisher (1921), the error on a correlation coefficient estimated from

$n$  independent samples is approximately proportional to  $1/\sqrt{n-3}$  when the populations from which those samples is drawn is uncorrelated. This means that we should expect  $\rho_{ij}$  to be a factor of  $\sim 2$  noisier for operational weather radar data than for our examples in sections 3 and 5. As explained in section 4, range sidelobe corrections are likely to be overestimated because of this extra noise. As a result we anticipate that the identification and correction of less prominent sidelobe artefacts will be more difficult for operational weather radars.

Future work will focus on the application of the flagging and correction procedures developed here to other meteorological scenarios to assess how well they perform. We also hope to use the methodology to characterise the sidelobes of our radar setup using different code and pulse repetition frequency options in order to best minimise the formation of sidelobes for the NCAS 35GHz radar at Chilbolton. Idealised simulations using synthetic echoes are also being performed to forward model the expected correlation coefficients which we will then measure using the real radar system.

#### *Acknowledgments.*

We would like to thank the staff at the Chilbolton Observatory for operating and maintaining the 35 GHz radar. Some of this work has been supported by the NCAS Facility for Ground-based Atmospheric Measurements (FGAM).

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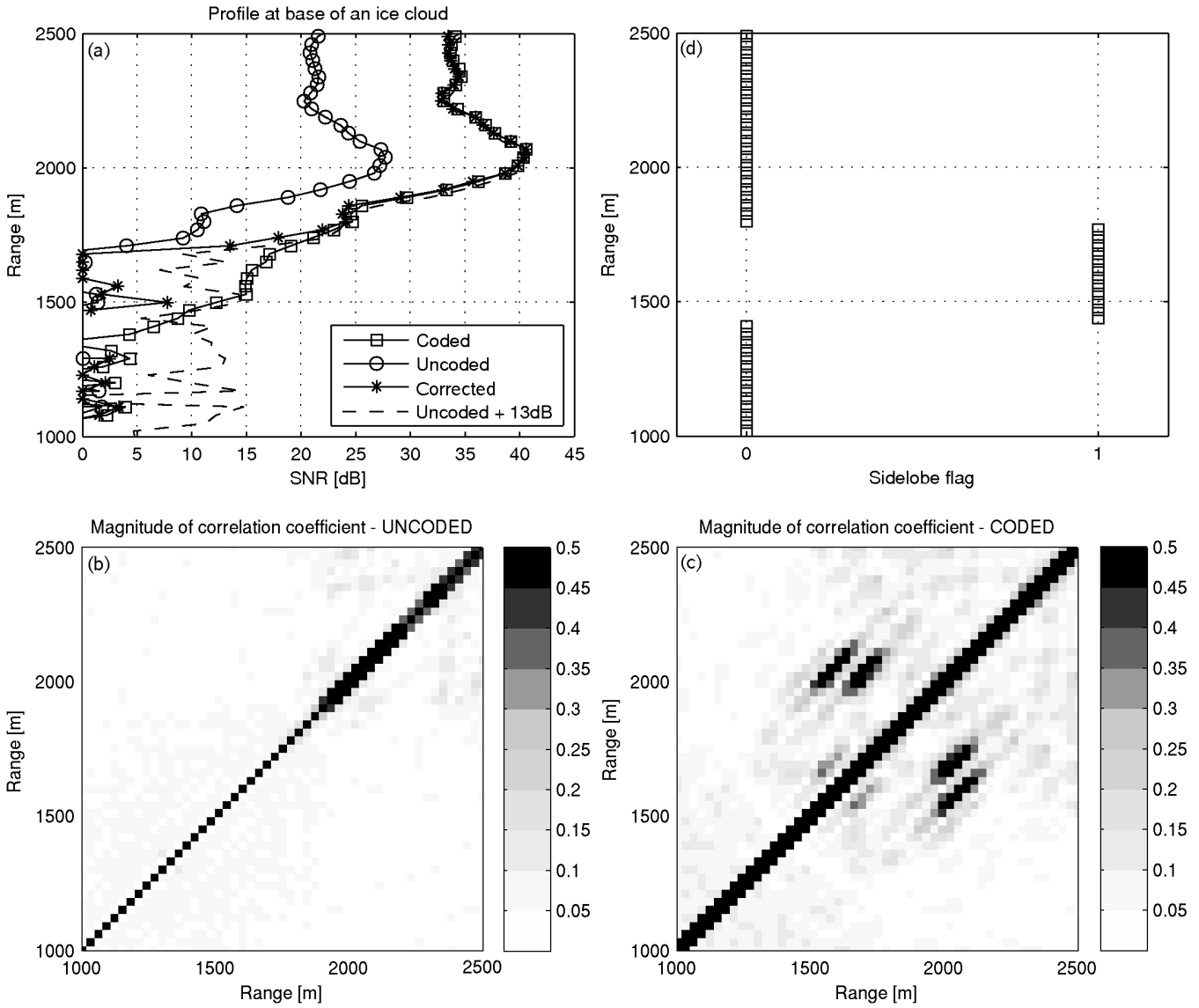


FIG. 4. Panel (a) shows the signal to noise profile at the base of a deep ice cloud using coded and uncoded pulses. Also shown is the uncoded profile with a 13dB offset, and a coded profile which has been corrected using the algorithm described in the text. Panels (b) and (c) show the magnitude of the correlation coefficient (grey scale) for pairs of range gates for uncoded (b) and coded (c) pulses. Panel (d) shows the result of the algorithm described in section 3b identifying sidelobe-corrupted data (value of 1), and unaffected data (value of 0).